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TITLE OF INVENTION

METHOD FOR FORMING OR DETECTING A SIGNAL SEQUENCE AND TRANSMITTER UNIT AND RECEIVER UNIT

APPLICANT(S) FOR DO/EO/US

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Applicant herewith submits to the United States Designated/Elected Office (DO/EO/US) the following items and other information:

1. ☒ This is a **FIRST** submission of items concerning a filing under 35 U.S.C. 371.
2. ☐ This is a **SECOND** or **SUBSEQUENT** submission of items concerning a filing under 35 U.S.C. 371.
3. ☒ This is an express request to promptly begin national examination procedures (35 U.S.C. 371(f)).
4. ☒ The US has been elected by the expiration of 19 months from the priority date (PCT Article 31).
5. ☒ A copy of the International Application as filed (35 U.S.C. 371(c)(2))
 - a. ☒ is attached hereto (required only if not communicated by the International Bureau).
 - b. ☐ has been communicated by the International Bureau.
 - c. ☐ is not required, as the application was filed in the United States Receiving Office (RO/US).
6. ☐ An English language translation of the International Application as filed (35 U.S.C. 371(c)(2)).
7. ☐ Amendments to the claims of the International Application under PCT Article 19 (35 U.S.C. 371(c)(3))
 - a. ☐ are attached hereto (required only if not communicated by the International Bureau).
 - b. ☐ have been communicated by the International Bureau.
 - c. ☐ have not been made; however, the time limit for making such amendments has NOT expired.
 - d. ☐ have not been made and will not be made.
8. ☐ An English language translation of amendments to the claims under PCT Article 19 (35 U.S.C. 371(c)(3)).
9. ☐ An oath or declaration of the inventor(s) (35 U.S.C. 371(c)(4)).
10. ☐ An English language translation of the annexes to the International Preliminary Examination Report under PCT Article 36 (35 U.S.C. 371(c)(5)).

Items 11 to 16 below concern other documents or information included:

11. ☐ An Information Disclosure Statement under 37 CFR 1.97 and 1.98.
12. ☐ An assignment document for recording. A separate cover sheet in compliance with 37 CFR 3.28 and 3.31 is included.
13. ☐ A FIRST preliminary amendment.
☐ A SECOND or SUBSEQUENT preliminary amendment.
14. ☐ A substitute specification.
15. ☐ A change of power of attorney and/or address letter.
16. ☒ Other items or information:

- ☒ Copy of annexes to the International Preliminary Examination Report
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Description

Method for forming and determining a signal sequence, a transmitting unit and a receiving unit

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The invention relates to a method for forming a signal sequence which is to be transmitted in particular for the purpose of synchronization of at least two transmission units, and to a method for determining this signal sequence, and corresponding transmitting and receiving units.

In signal transmission systems, such as mobile radio systems, it is necessary for one of the communication partners (a first transmission unit) to identify specific defined signals which are transmitted by another communication partner (second transmission unit). These may be, for example, so-called synchronization bursts (synchronization radio blocks) for synchronization of two synchronization partners, such as radio stations, or may be so-called access bursts.

In order to detect and to identify such received signals reliably against the background noise, it is known for the received signal to be continuously correlated with a predetermined signal sequence for a defined period of time, and for the correlation sum to be formed over the time period of the predetermined signal sequence. That region of the received signal which produces a maximum correlation sum corresponds to the sought signal. The synchronization signal from the base station in a digital mobile radio system is preceded, for example, by a signal sequence as a so-called training sequence, which is detected or determined in the manner just described in the mobile station, by correlation with the stored signal sequence. The mobile stations can thus be synchronized to the base station.

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Correlation calculations of such a type are also required in the base station, for example for Random Access Channel (RACH) detection. Furthermore, a correlation calculation is also carried out in order to
 5 establish the channel impulse response and the signal delay times of received signal bursts.

The correlation sum is in this case calculated as follows:

$$S_m = \sum_{i=0}^{n-1} E(i+m) * K(i)$$

10 where $E(i)$ is a received signal sequence derived from the received signal and $K(i)$ is the predetermined signal sequence, with i running from 0 to $n-1$. The correlation sum S_m is calculated successively for a number of signal sequences $E(i)$ at different times and
 15 obtained from the received signal, and the maximum value of S_m is then established. If k successive correlation sums are calculated, then the calculation effort comprises $k * n$ operations, with a multiplication and addition together being counted as
 20 one operation.

The calculation of the correlation sums is thus highly complex and, particularly for real-time applications such as voice communication or video telephony, or in CDMA systems, requires powerful and
 25 thus expensive processors, which consume a large amount of power during the calculation process. For example, a known signal sequence of length 256 chips (in CDMA, a transmitted bit is also referred to as a chip) has to be determined for synchronization of the UMTS mobile
 30 radio system, which is currently being standardized. The sequence is repeated every 2560 chips. Since the mobile station is initially operating asynchronously with respect to the chip clock, the received signal must be oversampled to ensure that an adequate signal
 35 is still obtained even if the sampling position is

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The invention is also based on the object of specifying methods and arrangements which allow signal sequences to be formed, and thus allow signal sequences to be specified, which can be determined easily in transmitted received signal sequences. The invention is also based on the object of specifying a method and arrangements which allow these signal sequences to be determined comparatively easily by the formation of correlation sums.

The object is achieved by the features in the independent patent claims. Developments can be found in the dependent claims.

The invention is based on the idea of forming signal sequences by repeating a second signal sequence element of length n_2 n_1 times and, in the process, modulating it with the first signal sequence element.

This allows signal sequences to be formed which, if they are contained in a received signal sequence, can be determined easily.

Specification of the method for forming signal sequences means that signal sequences which can be formed or are obtained by means of such a method are also within the scope of the invention. In particular, this also includes their use in data transmission systems, especially for the purpose of synchronization of a mobile station to a base station.

In order to determine a predetermined signal sequence, which is contained in a received signal sequence, by establishing correlation sums, a partial correlation sum sequence of the second signal sequence element is calculated with corresponding parts of the received signal sequence. In order to calculate a correlation sum, n_1 elements of the partial correlation sum sequence are selected and are multiplied by the first signal sequence element, in the sense of a scalar product.

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mobile radio network,

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invention,

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of a method for calculating a correlation sum,

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Figure 1 shows a cellular mobile radio network, such as the GSM (Global System for Mobile Communication) system, which comprises a large number of mobile switching centers MSC which are networked with one another and produce access to a landline network PSTN/ISDN. Furthermore, these mobile switching centers MSC are each connected to at least one base station controller BSC, which can also be formed by a data processing system. A similar architecture is also used in a UMTS (Universal Mobile Telecommunication System).

Each base station controller BSC is in turn connected to at least one base station BS. A base station BS such as this is a radio station which can set up a radio link via a radio interface to other radio stations, so-called mobile stations MS. Information can be transmitted within radio channels f , which are within frequency bands b , by means of radio signals between the mobile stations MS and the base station BS associated with these mobile stations MS. The range of the radio signals of a base station essentially defines a radio cell FZ.

Base stations BS and a base station controller BSC can be combined to form a base station system BSS. The base station system BSS is in this case also responsible for radio channel administration and assignment, data rate matching, monitoring of the radio transmission path, handover procedures and, in the case of a CDMA system, for allocation of the spread code set to be used, and transmits the signaling information required for this purpose to the mobile stations MS.

In the case of a duplex system, the frequency bands which are provided in FDD (Frequency Division Duplex) systems such as the GSM system for the uplink u (mobile station (transmitting unit) to the base station (receiving unit)) may be different to those provided for the downlink d (base station (transmitting unit) to

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the mobile station (receiving unit)). Within the
different frequency

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bands b , a number of frequency channels f can be provided by means of an FDMA (Frequency Division Multiple Access) method.

In the context of the present application, the term transmission unit also means a communication unit, transmitting unit, receiving unit, communications terminal, radio station, mobile station or base station. Terminology and examples which are used in the course of this application often also relate to a GSM mobile radio system; however, they are in no way limited to this system and, on the basis of the description, can also equally be applied by a person skilled in the art to other mobile radio systems, possibly future mobile radio systems, such as CDMA systems and, in particular, wideband CDMA systems.

Multiple access methods allow data to be transmitted efficiently via a radio interface, to be separated and to be allocated to one or more specific links and to the appropriate subscriber. A time division multiple access method TDMA, a frequency division multiple access method FDMA, a code division multiple access method CDMA or a combination of a number of these multiple access methods can be used for this purpose.

In the case of FDMA, the frequency band b is subdivided into a number of frequency channels f ; these frequency channels are split into timeslots ts by the time-division multiple access method TDMA. The signals transmitted within a timeslot TS and a frequency channel f can be separated by connection-specific spread codes, so-called CDMA codes cc , which are modulated onto the data.

The physical channels produced in this way are allocated to logical channels in accordance with a specified scheme. A fundamental distinction can be drawn between two types of logical channel: signaling channels (or control channels) for transmitting

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signaling information (or control information), and
traffic channels (TCH) for transmitting user data.

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The signaling channels are further subdivided into:

- Broadcast Channels
- Common Control Channels
- 5 - Dedicated/Access Control Channel (DCCH/ACCH)

The group of Broadcast Channels includes the Broadcast Control Channel BCCH by means of which the MS receives radio information from the Base Station System BSS, the Frequency Correction Channel FCCH and the Synchronization Channel SCH. The Common Control Channels include the Random Access Channel RACH. The radio blocks or signal sequences which are transmitted to provide these logical channels can in this case contain signal sequences $K(i)$, so-called correlation sequences, for various purposes, and signal sequences $K(i)$ can be transmitted on these logical channels for various purposes.

By way of example, a method for synchronization of a mobile station MS to a base station BS will be explained in the following text: during a first step of the initial base station search or cell search (initial cell search procedure), the mobile station uses the primary synchronization channel (SCH (PSC)) in order to achieve timeslot synchronization with the strongest base station. This can be ensured by a matched filter or an appropriate circuit which is matched to the primary synchronization code c_p but is transmitted by all the base stations. In this case, the same primary synchronization code c_p , of length 256, is transmitted by all the base stations BS.

The mobile station determines, by means of correlation from a received sequence, the received signal sequences $K(i)$ on the basis of a principle which is explained in Figures 6 to 11 and the associated description. In this case, peaks are emitted at the output of a matched filter for each received signal sequence of each base station which is located within

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the reception area of the mobile station. The detection of the position of the strongest peak makes it possible to determine the

| | 1980 | 1981 | 1982 | 1983 | 1984 | 1985 | 1986 | 1987 | 1988 | 1989 | 1990 | 1991 | 1992 | 1993 | 1994 | 1995 | 1996 | 1997 | 1998 | 1999 | 2000 | 2001 | 2002 | 2003 | 2004 | 2005 | 2006 | 2007 | 2008 | 2009 | 2010 | 2011 | 2012 | 2013 | 2014 | 2015 | 2016 | 2017 | 2018 | 2019 | 2020 | 2021 | 2022 | 2023 | 2024 | 2025 | 2026 | 2027 | 2028 | 2029 | 2030 | 2031 | 2032 | 2033 | 2034 | 2035 | 2036 | 2037 | 2038 | 2039 | 2040 | 2041 | 2042 | 2043 | 2044 | 2045 | 2046 | 2047 | 2048 | 2049 | 2050 | 2051 | 2052 | 2053 | 2054 | 2055 | 2056 | 2057 | 2058 | 2059 | 2060 | 2061 | 2062 | 2063 | 2064 | 2065 | 2066 | 2067 | 2068 | 2069 | 2070 | 2071 | 2072 | 2073 | 2074 | 2075 | 2076 | 2077 | 2078 | 2079 | 2080 | 2081 | 2082 | 2083 | 2084 | 2085 | 2086 | 2087 | 2088 | 2089 | 2090 | 2091 | 2092 | 2093 | 2094 | 2095 | 2096 | 2097 | 2098 | 2099 | 2100 | 2101 | 2102 | 2103 | 2104 | 2105 | 2106 | 2107 | 2108 | 2109 | 2110 | 2111 | 2112 | 2113 | 2114 | 2115 | 2116 | 2117 | 2118 | 2119 | 2120 | 2121 | 2122 | 2123 | 2124 | 2125 | 2126 | 2127 | 2128 | 2129 | 2130 | 2131 | 2132 | 2133 | 2134 | 2135 | 2136 | 2137 | 2138 | 2139 | 2140 | 2141 | 2142 | 2143 | 2144 | 2145 | 2146 | 2147 | 2148 | 2149 | 2150 | 2151 | 2152 | 2153 | 2154 | 2155 | 2156 | 2157 | 2158 | 2159 | 2160 | 2161 | 2162 | 2163 | 2164 | 2165 | 2166 | 2167 | 2168 | 2169 | 2170 | 2171 | 2172 | 2173 | 2174 | 2175 | 2176 | 2177 | 2178 | 2179 | 2180 | 2181 | 2182 | 2183 | 2184 | 2185 | 2186 | 2187 | 2188 | 2189 | 2190 | 2191 | 2192 | 2193 | 2194 | 2195 | 2196 | 2197 | 2198 | 2199 | 2200 | 2201 | 2202 | 2203 | 2204 | 2205 | 2206 | 2207 | 2208 | 2209 | 2210 | 2211 | 2212 | 2213 | 2214 | 2215 | 2216 | 2217 | 2218 | 2219 | 2220 | 2221 | 2222 | 2223 | 2224 | 2225 | 2226 | 2227 | 2228 | 2229 | 2230 | 2231 | 2232 | 2233 | 2234 | 2235 | 2236 | 2237 | 2238 | 2239 | 2240 | 2241 | 2242 | 2243 | 2244 | 2245 | 2246 | 2247 | 2248 | 2249 | 2250 | 2251 | 2252 | 2253 | 2254 | 2255 | 2256 | 2257 | 2258 | 2259 | 2260 | 2261 | 2262 | 2263 | 2264 | 2265 | 2266 | 2267 | 2268 | 2269 | 2270 | 2271 | 2272 | 2273 | 2274 | 2275 | 2276 | 2277 | 2278 | 2279 | 2280 | 2281 | 2282 | 2283 | 2284 | 2285 | 2286 | 2287 | 2288 | 2289 | 2290 | 2291 | 2292 | 2293 | 2294 | 2295 | 2296 | 2297 | 2298 | 2299 | 2300 | 2301 | 2302 | 2303 | 2304 | 2305 | 2306 | 2307 | 2308 | 2309 | 2310 | 2311 | 2312 | 2313 | 2314 | 2315 | 2316 | 2317 | 2318 | 2319 | 2320 | 2321 | 2322 | 2323 | 2324 | 2325 | 2326 | 2327 | 2328 | 2329 | 2330 | 2331 | 2332 | 2333 | 2334 | 2335 | 2336 | 2337 | 2338 | 2339 | 2340 | 2341 | 2342 | 2343 | 2344 | 2345 | 2346 | 2347 | 2348 | 2349 | 2350 | 2351 | 2352 | 2353 | 2354 | 2355 | 2356 | 2357 | 2358 | 2359 | 2360 | 2361 | 2362 | 2363 | 2364 | 2365 | 2366 | 2367 | 2368 | 2369 | 2370 | 2371 | 2372 | 2373 | 2374 | 2375 | 2376 | 2377 | 2378 | 2379 | 2380 | 2381 | 2382 | 2383 | 2384 | 2385 | 2386 | 2387 | 2388 | 2389 | 2390 | 2391 | 2392 | 2393 | 2394 | 2395 | 2396 | 2397 | 2398 | 2399 | 2400 | 2401 | 2402 | 2403 | 2404 | 2405 | 2406 | 2407 | 2408 | 2409 | 2410 | 2411 | 2412 | 2413 | 2414 | 2415 | 2416 | 2417 | 2418 | 2419 | 2420 | 2421 | 2422 | 2423 | 2424 | 2425 | 2426 | 2427 | 2428 | 2429 | 2430 | 2431 | 2432 | 2 |
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timing of the strongest base station modulo the slot length. In order to ensure greater reliability, the output from the matched filter can be accumulated non-coherently over the number of timeslots. The mobile station thus carries out a correlation over a signal sequence of length 256 chips as a matched filter operation.

The synchronization code cp is in this case formed, or can be formed or is obtained in such a way, corresponding to a signal sequence $K(i)$ based on a principle which is explained in Figure 5 and the associated description. The signal sequence $K(i)$ or the synchronization code cp of length 256 is in this case formed, or can be formed in such a way, from two signal sequence elements $K1(j)$, $K2(k)$, which each have a length of 16. These signal sequence elements in this case form a signal sequence element pair $(K1(j); K2(k))$.

A signal sequence $K(i)$ which is obtained in such a manner can in this case also be referred to as a "hierarchical signal sequence". A signal sequence element can also be referred to as "short correlation sequence".

Figure 2 shows a radio station, which may be a mobile station MS, comprising a control unit or interface unit MMI, a control device STE, a processing device VE, a power supply device SE, a receiving device EE and, possibly, a transmitting device SE.

The control device STE essentially comprises a programmable microcontroller MC which can access memory modules SPE for writing and reading. The microcontroller MC controls and monitors all the major elements and functions of the radio station.

The processing device VE may also be formed by a digital signal processor DSP, which can likewise access memory modules SPE. The processing

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device VE can also provide addition and multiplication means.

The program data which are required for controlling the radio station and the communication sequence, in particular the signaling procedures as well, and information obtained during the processing of signals are stored in the volatile or non-volatile memory modules (SPE). In addition, signal sequences $K(i)$ which are used for correlation purposes, and intermediate results from correlation sum calculations, can be stored in these memory modules SPE. The signal sequences $K(i)$ which are within the scope of the invention may thus be stored in the mobile station and/or in the base station.

It is also possible for one or more signal sequence elements or signal sequence element pairs ($K1(j)$; $K2(k)$) to be stored in the mobile station and/or in the base station. It is also possible for a signal sequence $K(i)$ to be formed from a signal sequence element pair ($K1(j)$; $K2(k)$) in the mobile station and/or in the base station.

In particular, a signal sequence $K(i)$ which is transmitted at fixed or variable intervals for synchronization purposes can be stored in a base station or in all the base stations in a system. The signal sequence element pair ($K1(j)$; $K2(k)$) from which the signal sequence $K(i)$ which is stored in the base station can be formed is stored in the base station MS and is used for synchronization of the mobile station to a base station in order to carry out correlation sum calculations with little computation complexity.

The signal sequences and the signal sequence elements can also be stored by storing appropriate information in any desired coded form and by means for storage such as volatile and/or non-volatile memory modules, or by means of appropriately configured adder or multiplier inputs or corresponding

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equivalent hardware configurations.

The radio frequency section HF may comprise the transmitting device SE, with a modulator and an amplifier V, and a receiving device EE with a demodulator and, likewise, with an amplifier. The analog audio signals and the analog signals originating from the receiving device EE are converted into digital signals by analog/digital conversion, and are processed by the digital signal processor DSP. After processing, the digital signals may be converted by digital/analog conversion into analog audio signals or other output signals, and analog signals to be supplied to the transmitting device SE. Modulation and demodulation may be carried out for this purpose.

The transmitting device SE and the receiving device EE are supplied via the synthesizer SYN with the frequency from a voltage controlled oscillator VCO. The voltage controlled oscillator VCO also allows the system clock to be produced to provide the clock for the processor devices in the radio station.

An antenna device ANT is provided for receiving and transmitting signals via the air interface of a mobile radio system. In some known mobile radio systems, such as the GSM (Global System for Mobile Communication), the signals are received and transmitted pulsed with respect to time, in so-called bursts.

The radio station may also be a base station BS. In this case, the loudspeaker element and the microphone element of the control unit MMI are replaced by a link to a mobile radio network, for example via a base station controller BSC and/or a switching center MSC. In order to interchange data simultaneously with a number of mobile stations MS, the base station BS has an appropriately large number of transmitting and receiving devices.

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Figure 3 shows a received signal sequence $E(1)$ which may be a signal sequence derived from a received signal and has a length w . In order to calculate a first correlation sum S_0 in accordance with the formula
 5 quoted initially, elements in a first section of this received signal sequence $E(1)$ are multiplied in pairs by the corresponding elements in the signal sequence $K(i)$ which has a length n , and the length of the resultant result element are added up to form the
 10 correlation sum S_0 .

In order to calculate a further correlation sum S_1 , the signal sequence $K(i)$ is shifted by one element to the right, as illustrated in graphical form in the figure, and the elements in the signal sequence $K(i)$
 15 are multiplied in pairs by the corresponding elements in the signal sequence $E(1)$, with the correlation sum S_1 once again being formed by summation of the resultant result elements.

The multiplication of the elements in the
 20 signal sequence in pairs by corresponding elements in the received signal sequence and subsequent summation can also be described vectorially as the formation of a scalar product, if the elements in the signal sequence and the elements in the received signal sequence are in
 25 each case combined to form a vector in a Cartesian coordinate system:

$$S_0 = \begin{pmatrix} K(0) \\ \vdots \\ K(i) \\ \vdots \\ K(n-1) \end{pmatrix} * \begin{pmatrix} E(0) \\ \vdots \\ E(i) \\ \vdots \\ E(n-1) \end{pmatrix}$$

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$$S1 = \begin{pmatrix} K(0) \\ \vdots \\ K(i) \\ \vdots \\ K(n-1) \end{pmatrix} * \begin{pmatrix} E(1) \\ \vdots \\ E(i+1) \\ \vdots \\ E(n) \end{pmatrix}$$

In the correlation sums S determined in this way, it is possible to search for the maximum, to compare the maximum of the correlation sums S with a predetermined threshold value, and in this way to determine whether the received signal $E(1)$ contains the predetermined signal sequence $K(i)$ and, if yes, where it is located in the received signal $E(1)$, and thus to synchronize two radio stations to one another and to detect data onto which an individual spread code has been modulated in the form of a signal sequence $K(i)$.

Figure 4 once again shows the received signal sequence $E(1)$ and, as a correlation sequence, a signal sequence $K(i)$, which is based on the signal sequence elements $K1(j)$, $K2(k)$.

Figure 5 shows the formation of a signal sequence $K(i)$ which is based on two signal sequence elements $K2(k)$ of length $n2$ and $K1(j)$ of length $n1$. For this purpose, the signal sequence element $K2(k)$ is repeated $n1$ times, with the signal sequence element $K1(j)$ being modulated onto it in the process. The formation of the signal sequence $K(i)$ can also be expressed mathematically by the following formula:

$$K(i) = K2(i \bmod n2) * K1(i \operatorname{div} n2), \text{ for } i = 0 \dots n1 * n2 - 1$$

This corresponds to the situation where $n1 = n2$ in the following relationship:

$$K(i) = K2(i \bmod n1) * K1(i \operatorname{div} n2).$$

In this case, \bmod denotes the integer remainder of a division, and div denotes the integer result of a division.

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This is represented in graphical form by a sequence f2 which comprises repeatedly, successively mapped signal sequence elements $K2(k)$, and a sequence f1, which is mapped by an expanded signal sequence element $K1(j)$ over the sequence f2.

The new signal sequence $K(i)$, of length n, is obtained by multiplication of the elements in the sequence f2 by the corresponding elements in the sequence f1 which are mapped over the sequence f2. This production of a signal sequence $K(i)$ is illustrated once again below in the figure, based on an example of two binary signal sequence elements of length 4.

The invention is not, of course, limited to signal sequence elements of length 4, or to signal sequences of length 16. The invention is also not limited to the mathematical description used above.

By way of example, the contents of the following illustration for signal sequence elements of length 16 and signal sequences of length 256 corresponds to the mathematical representation used above, and is thus likewise included in the invention: a is a signal sequence element with a length of 16

$a = \langle x_1, x_2, \dots, x_{16} \rangle;$

the signal sequence y of length 256 is generated by repeating the signal sequence element a 16 times, with a having a second signal sequence element of length 16 modulated onto it:

$y = \langle a, a, a, \underline{\underline{a}}, \underline{\underline{a}}, a, \underline{\underline{a}}, \underline{\underline{a}}, a, a, a, \underline{\underline{a}}, a, \underline{\underline{a}}, a, a \rangle$

where the double underscore indicates the second signal sequence element and the modulation by the second signal sequence element.

The signal sequence of length 256 formed in this way can, for example, be transmitted for synchronization purposes as a primary synchronization code cp of length 256.

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Signal sequences $K(i)$ formed in this way can be used to simplify the calculation of correlation sums of these signal sequences $K(i)$ with received signal sequences $E(l)$.

5 A schematic illustration of such a simplified process for calculating correlation sums S , which is thus also faster and involves less effort, is illustrated in Figures 6 to 8, which are referred to in the following text.

10 A partial correlation sum $TS(z)$ is formed first of all. By way of example, the correlation sum of the second signal sequence element $K2(k)$ with the corresponding section of the received signal sequence $E(l)$ is formed for this purpose for the first element
15 in the partial correlation sum sequence $TS(0)$.

$$TS(0) = \begin{pmatrix} K2(0) \\ \vdots \\ K2(k) \\ \vdots \\ K2(n2-1) \end{pmatrix} \cdot \begin{pmatrix} E(0) \\ \vdots \\ E(k) \\ \vdots \\ E(n2-1) \end{pmatrix}$$

As illustrated in graphical form, the second signal sequence element $K2(k)$ is shifted by one element for the second element in the partial correlation sum
20 sequence $TS(1)$, and the correlation sum is likewise formed with the corresponding element in the received signal sequence $E(l)$, etc.

$$TS(1) = \begin{pmatrix} K2(0) \\ \vdots \\ K2(k) \\ \vdots \\ K2(n2-1) \end{pmatrix} \cdot \begin{pmatrix} E(1) \\ \vdots \\ E(k+1) \\ \vdots \\ E(n2) \end{pmatrix}$$

25 The n -th element in the partial correlation sum sequence $TS(n1 \cdot n2 - 1)$ is calculated in a corresponding manner after $n-1$ shifts of the second signal sequence element $K2(k)$ with respect to the received signal sequence $E(l)$.

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$$TS(n-1) = \begin{pmatrix} K2(0) \\ \vdots \\ K2(k) \\ \vdots \\ K2(n2-1) \end{pmatrix} * \begin{pmatrix} E(n-1) \\ \vdots \\ E(k+n-1) \\ \vdots \\ E(n2+n-2) \end{pmatrix}$$

The partial correlation sum sequence $TS(z)$ produced in this way is illustrated in the upper part of Figure 7. Each $n2$ -th element is now selected from this partial correlation sum sequence and is multiplied in pairs by the corresponding element in the first signal sequence element $K1(j)$.

If the selected elements in the partial correlation sum sequence $TS(z)$ and the first signal sequence element $K1(j)$ are respectively combined to form vectors, then the first correlation sum $S0$ is produced by the scalar product of these two vectors.

$$S0 = \begin{pmatrix} K1(0) \\ \vdots \\ K1(j) \\ \vdots \\ K1(n1-1) \end{pmatrix} * \begin{pmatrix} TS(0) \\ \vdots \\ TS(j*n2-1) \\ \vdots \\ TS((n1-1)*n2-1) \end{pmatrix} =$$

The lower part of Figure 7 shows the corresponding calculation for further correlation sums $S1$ and $S2$, respectively, by the selection of the $n2$ -th elements which have been shifted by one or two places to the right from those elements which were selected first:

$$S1 = \begin{pmatrix} K1(0) \\ \vdots \\ K1(j) \\ \vdots \\ K1(n1-1) \end{pmatrix} * \begin{pmatrix} TS(1) \\ \vdots \\ TS(j*n2) \\ \vdots \\ TS((n1-1)*n2) \end{pmatrix}$$

The storage of the partial correlation sums TS once they have been calculated allows them to be referred back to when further correlation sums are calculated later, and thus makes it possible to
 5 dispense with the corresponding computation steps.

Depending on the design variant, it is either possible first of all to calculate the complete partial correlation sum sequence TS(z) over the entire received signal sequence E(1) and then to calculate the
 10 individual correlation sums, or to calculate the corresponding additionally required partial correlation sums only when required in order to calculate a new correlation sum.

Figure 8 once again shows the method, which
 15 comprises two steps, for calculating correlation sums S, on this occasion with reference to the example, illustrated in Figure 5, of two binary signal sequence elements of length 4.

In a first step, the partial correlation sums
 20 TS(z) of the second signal sequence element K2(k) ++++ are calculated with corresponding sections of the received signal sequence E(1) and then, in a second step, every fourth element in the partial correlation sum sequence TS(z) produced in this way is selected, is
 25 multiplied by the corresponding element in the first signal sequence element K1(j) +---+, and these are added up to form the correlation sequence S0.

The thick lines in this case represent those calculation steps which need to be carried out newly in
 30 order to calculate a further correlation sum S1 in the situation where the other partial correlation sums TS have already been calculated and stored in advance.

This design variant can be carried out using memory as efficiently as possible if every n2-th
 35 partial correlation sum is calculated first of all. The sample values are buffer stored for this purpose.

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Figures 9 and 10 illustrate another design variant to simplify the calculation of correlation sums S on the basis of the example, already mentioned above, of two binary signal sequence elements of length 4.

5 In this case, every fourth element in the received signal sequence $E(l)$ is first of all selected, and the partial correlation sum sequence $TS(z)$ of the elements selected in this way with the signal sequence element $K1(j)$ is formed. Four successive elements are
10 in each case selected from the partial correlation sum sequence $TS(z)$ obtained in this way and are multiplied in pairs by corresponding elements in the signal sequence element $K2(k)$ and the resultant result elements are added up to form the correlation sum S . In
15 this case, the thick lines once again represent the additionally required steps to calculate a further correlation sum $S1$ in the situation where the other partial correlation sums TS have already been calculated and stored in advance.

20 Figure 10 once again shows the calculation of a first correlation sum $S0$ in which every fourth element in the received signal sequence $E(l)$ is first of all selected, these elements are multiplied by
25 corresponding elements in the first signal sequence element $K1(j) +---+$, and the partial correlation sum $TS(0)$ is calculated by summation of the result elements. In a second step, the first four successive elements in the partial correlation sum sequence $TS(z)$ are multiplied by the corresponding elements in the
30 second signal sequence element $K2(k) ++--$, and the resultant result elements are added up to form the correlation sum $S0$.

In this design variant, less memory is required for buffer storage of the partial correlation sums, if
35 the sums are calculated successively.

In one preferred design variant of the invention, Barker sequences are used as signal sequence

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elements since, with respect to their length, these
offer best-possible autocorrelation

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characteristics. The use of a signal sequence of length 169 is particularly suitable for the abovementioned methods, and is formed by two Barker signal sequence elements of length 13. Such a sequence can
5 advantageously be used in particular in a UMTS mobile radio system.

Depending on the design variant, the elements in the signal sequence elements may assume values on the unit circle, any given real values, or any given
10 complex values.

Furthermore, a further refinement of the invention provides for two signal sequence elements of the same length to be used. It is also possible to use the same sequence for both signal sequence elements. In
15 addition, it may be advantageous to use the mirror image of the first signal sequence element as the second signal sequence element: $K1(j) = K2(n1-j)$.

It is also possible to use a shortened signal sequence with less than $n1 * n2$ values. In this case,
20 the last element in the partial correlation sum sequence (for calculating a new correlation sum) is first of all calculated in shortened form in order to calculate the correlation sums, and is then calculated completely (in order to calculate the partial
25 correlation sum sequence required for the following correlation sums). This allows the generation of signal sequences of any desired length.

Thus, using the nomenclature introduced above, a shortened signal sequence $Ks(i)$ is used, whose length
30 ns is less than $n1*n2$. $ns = n1*n2 - \delta$. $n1$ can be selected such that $\delta < n2$.

The calculation is carried out analogously to the method already described, just with shortened partial correlation sequences Tss additionally being
35 calculated, whose length is likewise δ shorter than $n2$.

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$$TSs(n-1) = \begin{pmatrix} K2(0) \\ \vdots \\ K2(k) \\ \vdots \\ K2(n2-1-\delta) \end{pmatrix} * \begin{pmatrix} E(n-1) \\ \vdots \\ E(k+n-1) \\ \vdots \\ E(n2+n-2-\delta) \end{pmatrix} =$$

The (shortened) first correlation sum $S0$ can then be calculated with the shortened partial correlation sum being used for the last term, in contrast to the method described above.

$$S0 = K1(0) * TS(0) + \dots + K2(k) * E(k+n-1) + \dots + K2(n2-1-\delta) * E(n2+n-2-\delta)$$

The unshortened partial correlation sum $TS(i)$ which will be required later is advantageously calculated using the stored partial correlation sequence $TSs(i)$.

It is also possible to use a lengthened signal sequence, that is to say with more than $n1 * n2$ values. In this case, the correlation sums are calculated by calculating the first part ($n1 * n2$ terms) using the above method, with additional terms also being added. The correlation sequence $K1(i)$ to this end contains a correlation sequence $K(i)$ according to the invention, but lengthened by additional elements. This method once again allows the generation of sequences of any desired length.

Thus, using the nomenclature introduced above, a lengthened correlation sequence $K1(i)$ is used, whose length is $n1$ greater than $n1 * n2$. Thus $K1(i) = K(i)$ for $0 \leq i \leq n1 * n2 - 1$.

The calculation is carried out analogously to the already described method, just with the additional terms also being calculated using any desired method, either conventionally or likewise according to the invention.

$$SO = K1(0) * TS(0) + \dots + K1(j * n2 - 1) * TS((n1 - 1) * n2 - 1) + \\ + K1(n1) * E((n1 - 1) * n2) + \dots + K1(n1) * E(n1)$$

In a further variant of the most recently
 5 described exemplary embodiment, additional values are
 inserted at the start and/or between the (modulated)
 repetitions of the signal sequence element K2. the
 elements which are not inserted are then processed
 further using a method as above, and the inserted
 10 elements are processed further either conventionally,
 or likewise using a method according to the invention.

Another development of the invention envisages
 the use of more than two signal sequence elements, with
 a signal sequence element itself comprising signal
 15 sequence elements.

A further refinement of the invention makes use
 of the regular (virtually periodic) structure of the
 aperiodic autocorrelation function of this signal
 sequence resulting from the regular construction
 20 principle of the signal sequence K(i). This means that
 the search for a signal produces not only a main
 maximum but also secondary maxima at regular intervals.
 The regularity of the position of the maxima can be
 utilized to speed up the search for the signal sequence
 25 in the received signal sequence. As soon as a secondary
 maximum has been found, the position of the other
 maxima can be predicted on the basis of the
 periodicity, that is to say the correlation sum is
 calculated only at these points. This allows the main
 30 maximum to be detected quickly. However, the supposed
 secondary maximum may also just be a value which has
 been increased randomly (owing to the amount of noise).
 In this case, no maximum will actually be found at the
 potential points for the expected main maximum. In this
 35 case, the hypothesis is thus rejected, and the
 calculation is continued conventionally.

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However, the regularity of the secondary maxima resulting from the construction principle of the signal sequences can also be used to

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eliminate and correct for interference secondary maxima in the correlation result. After detection of the maximum, the secondary maxima can be calculated from this maximum, and this value can be subtracted from the corresponding correlation results. This results in the correlation result for a (hypothetical) sequence with a perfect autocorrelation function. The regularity of the secondary maxima thus results in a highly simplified calculation.

10 The invention is not limited to radio transmission systems, but can also be used for other transmission methods, for example acoustic methods (ultrasound), in particular for sonography purposes, or optical methods, for example infrared measurements
15 based on Lidar principles. One further field of application is the investigation of changes in the spectral composition of back-scattered signals.

 The formation of signal sequences, their transmission and the calculation of correlation sums
20 for these signal sequences with received signal sequences may be used in various technical fields:

 - for the purpose of synchronization of two transmission units, for example radio stations, in particular the use of these sequences in the
25 synchronization channel in CDMA mobile radio systems, such as the UMTS system which is currently being standardized,

 - in data transmission by means of transmitted symbols spread by means of the signal sequence or data
30 in spread spectrum systems, in particular for determining transmitted symbols and data onto which such a signal form has been modulated,

 - in instrumentation, for distance and object measurement,

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5 - for establishing transmission characteristics of the transmission channel located between transmission units, such as a transmitting unit and a receiving unit, in radar technology, in order to establish the position and orientation of an object and/or further parameters which are dependent on the geometry and the specific reflection characteristics of the object,

10 - to establish transmission characteristics of the transmission channel which is located between the transmitter and receiver, in radar technology for determining parameters of a back-scattering medium, in particular the ionosphere, in particular by incoherent scattering,

15 - for establishing transmission characteristics of the transmission channel which is located between transmission units, such as a transmitting unit and receiving unit, in particular for establishing multipath propagation in instrumentation or
20 communications technology. In this case, the way in which the propagation characteristics of the transmission channel (channel impulse response) changes with time is determined during communication by means of the correlation result. In particular, additional
25 multipath propagation paths are determined. For this purpose, the signal sequences $K(i)$ may also be transmitted in the form of a midamble within a radio block. This knowledge can then be used further in an otherwise conventional receiving unit.

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Patent Claims

1. A method for synchronization of a base station (BS) to a mobile station (MS),
 - 5 - in which the base station transmits a signal sequence $K(i)$ of length n , which can be formed in such a way that [lacuna],
 - in which a second signal sequence element $K2(k)$ of length $n2$ is repeated $n1$ times and, in the process, has
 - 10 a first signal sequence element $K1(j)$ of length $n1$ modulated onto it,
 - in which $n1$ is equal to $n2$, and
 - in which this signal sequence $K(i)$ is determined in a mobile station.
- 15 2. The method as claimed in claim 1, in which n is equal to 256, $n1$ is equal to 16, and $n2$ is equal to 16.
3. The method as claimed in one of the preceding claims, in which
- 20 the signal sequence $K(i)$ is formed by modulation of the second signal sequence element $K2(k)$ in accordance with the following rule: $K(i) = K2(i \bmod n2) * K1(i \div n2)$.
4. The method as claimed in one of the preceding claims,
- 25 - in which the predetermined signal sequence $K(i)$ contained in a received signal sequence $E(l)$ is determined in the mobile station by establishing the correlation sums S of the signal sequence $K(i)$ with corresponding sections of the received signal sequence
- 30 $E(l)$, with
 - a partial correlation sum sequence $TS(z)$ of the signal sequence element $K2(k)$ with corresponding parts of the received signal sequence $E(l)$ being calculated, and
 - 35 - $n1$ elements of the partial correlation sum sequence $TS(z)$ being selected in order to calculate a correlation sum S , and being multiplied by the signal sequence element $K1(j)$ in the sense of a scalar product.

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5. The method as claimed in claim 4, in which

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n1 in each case n2-th elements of the partial correlation sum sequence TS(z) are selected in order to calculate a correlation sum S.

6. The method as claimed in one of claims 1, 2 or 3,

- in which the predetermined signal sequence K(i) contained in a received signal sequence E(l) is determined in the mobile station by establishing the correlation sums S of the signal sequence K(i) with corresponding sections of the received signal sequence E(l), with

- a partial correlation sum sequence TS(z) of the signal sequence element K1(j) with selected elements of the received signal sequence E(l) being calculated, and
- n2 elements of the partial correlation sum sequence TS(z) being multiplied by the signal sequence element K2(k), in the sense of a scalar product, in order to calculate a correlation sum S.

7. The method as claimed in claim 6, in which

n1 in each case n2-th elements of the received signal sequence E(l) are selected in order to calculate a partial correlation sum TS.

8. The method as claimed in one of claims 4 to 7, in which calculated partial correlation sums TS are stored, and are used in order to calculate a further correlation sum S.

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Abstract

Method for forming and determining a signal sequence, a transmitting unit and a receiving unit

5

Formation of signal sequences which are based on signal sequence elements, with the second signal sequence element being repeated and in the process being modulated with the first signal sequence element.

10 Use of these signal sequence elements for synchronization of two transmission units in order to simplify the calculation of correlation sums in a two-stage calculation method, with a partial correlation sum sequence being calculated first of all.

15

Figure 5

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FIG 1

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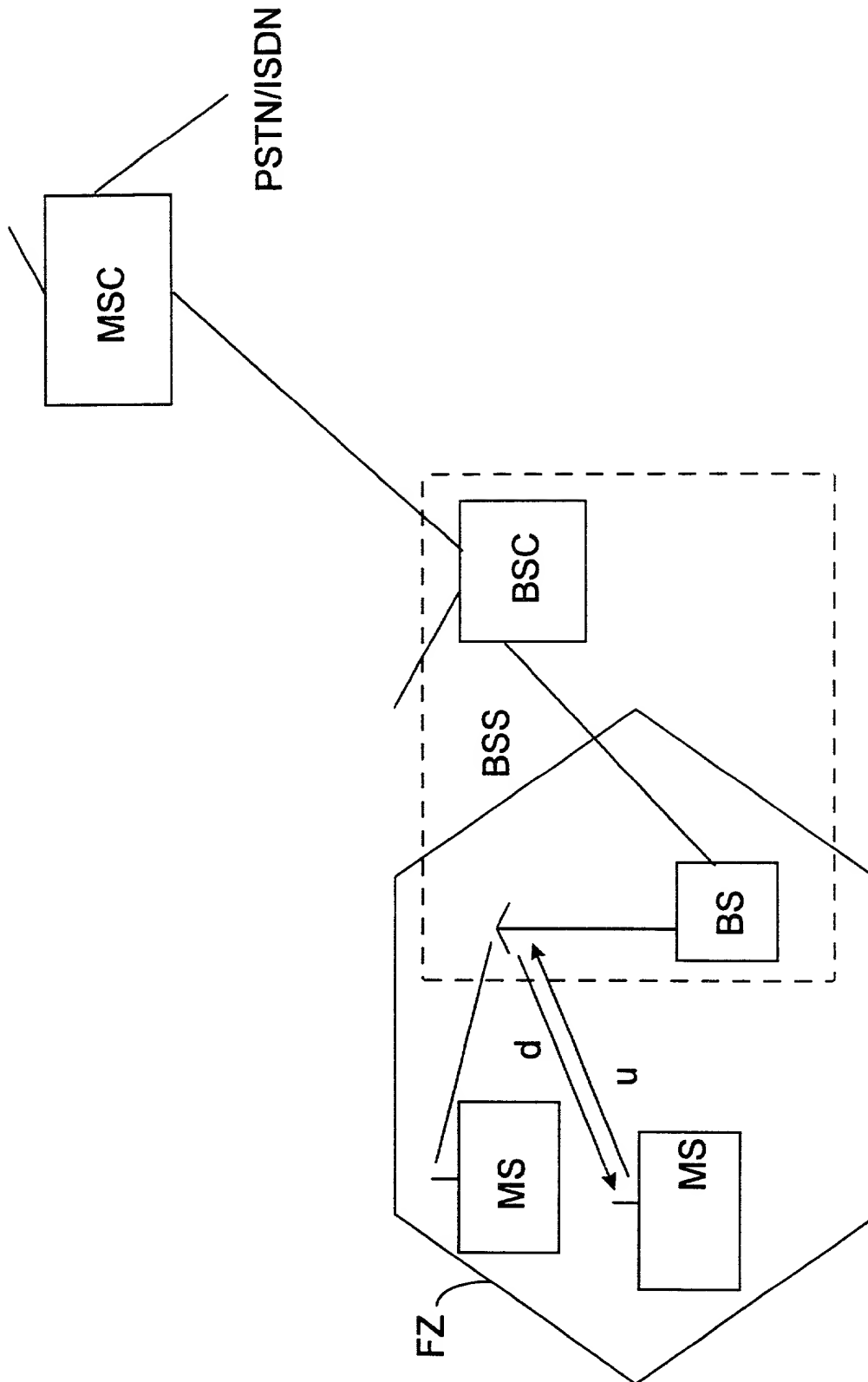


FIG 2

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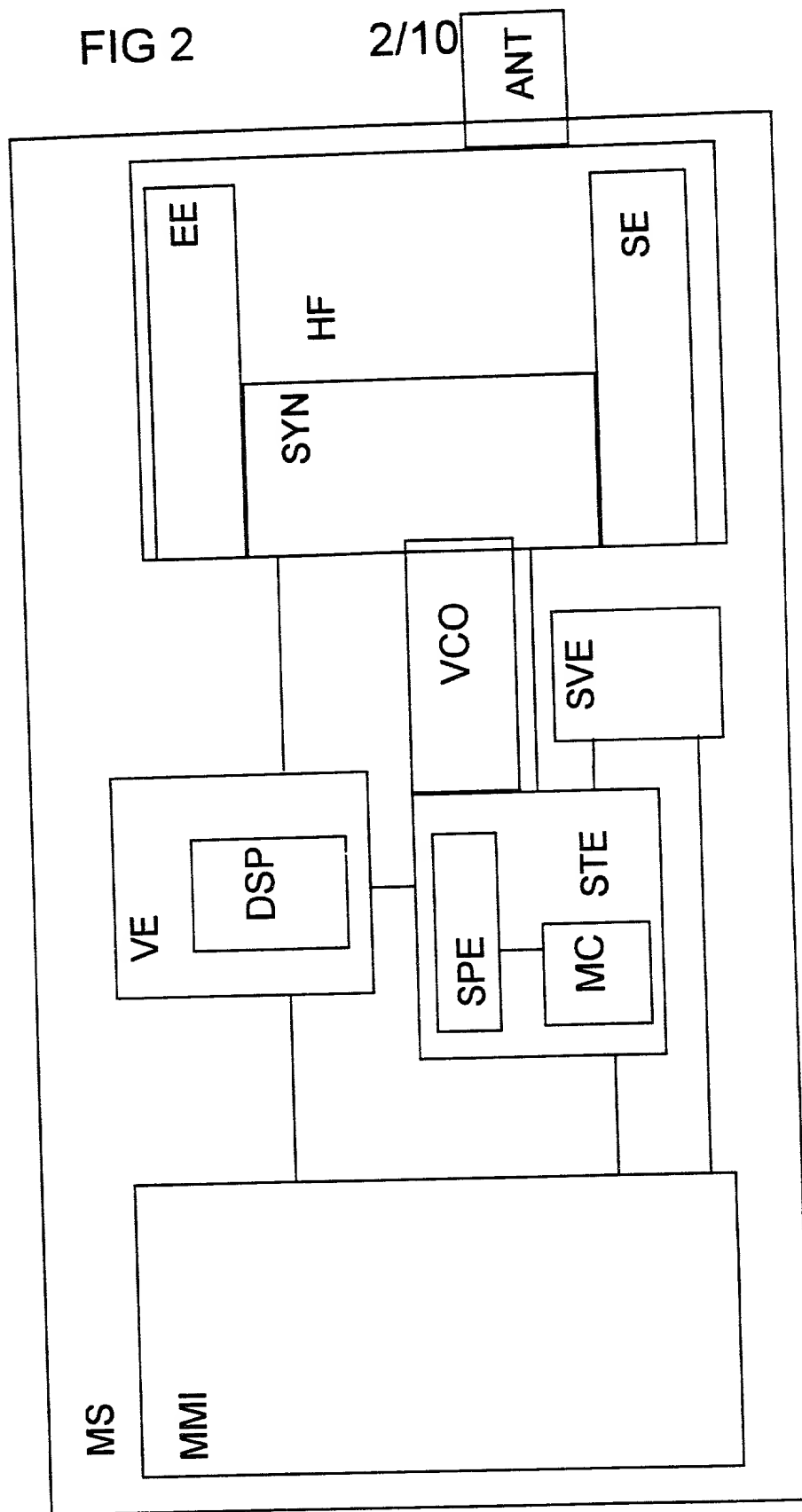
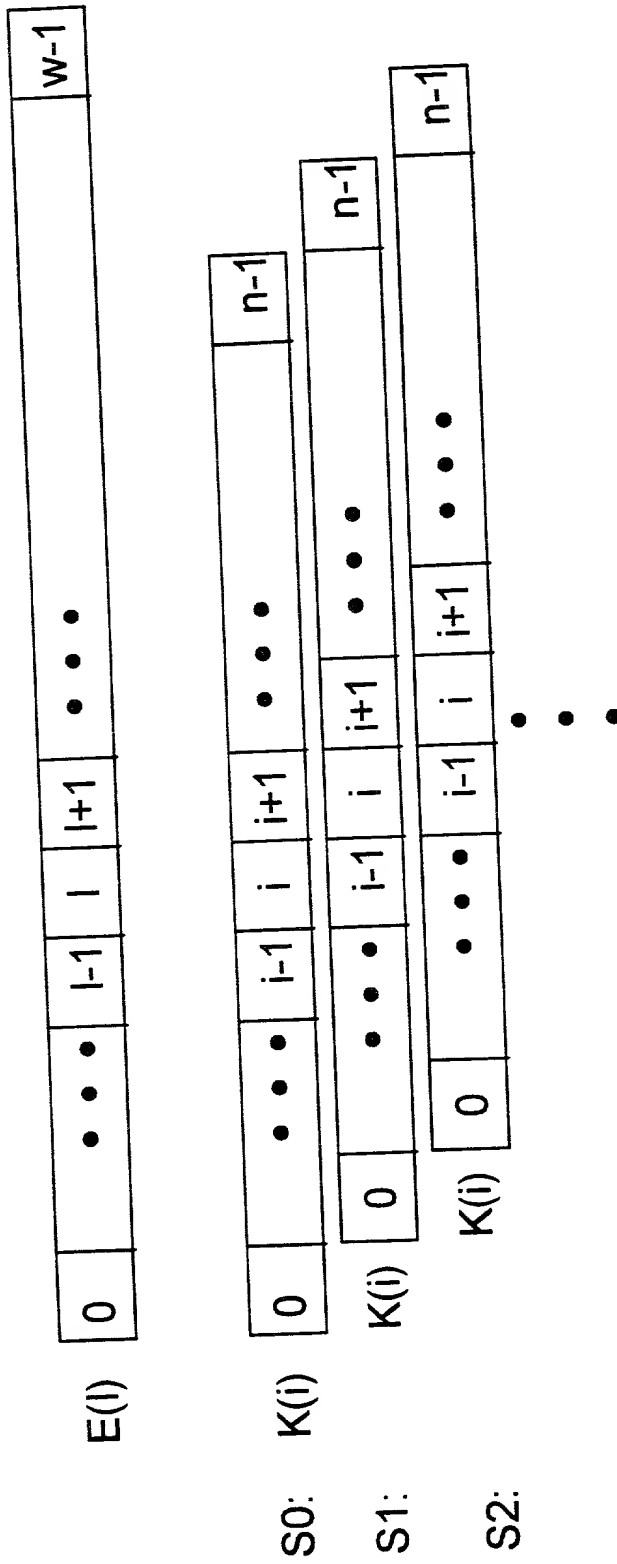
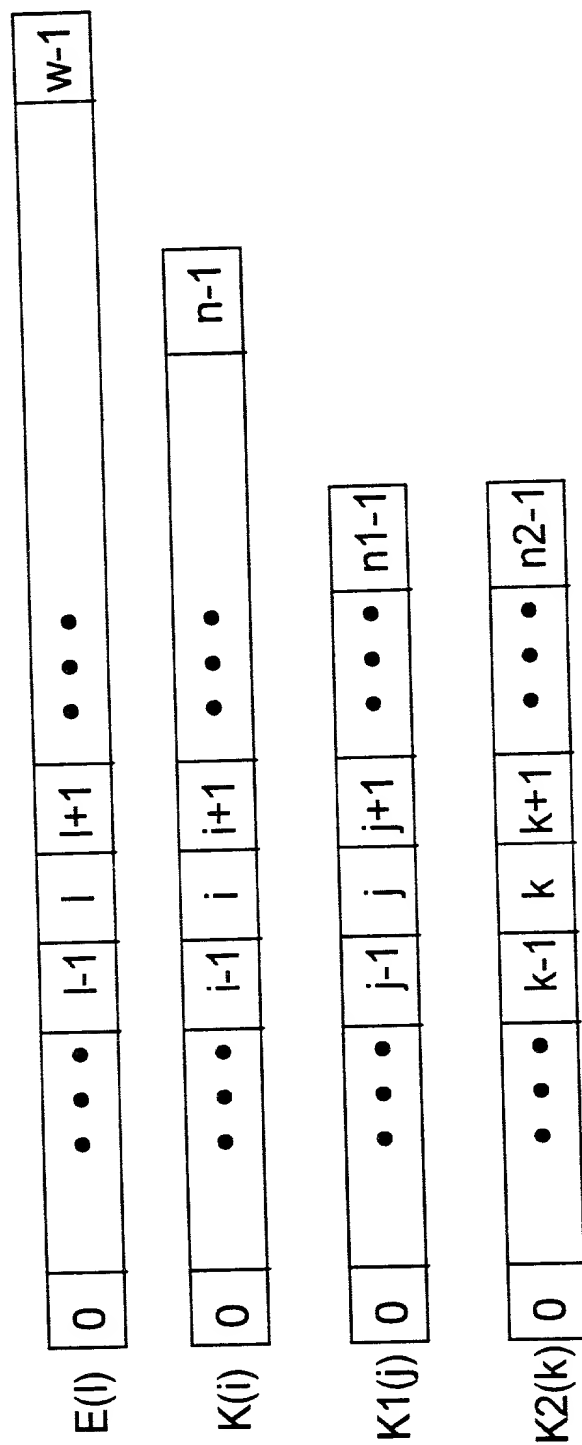


FIG 3

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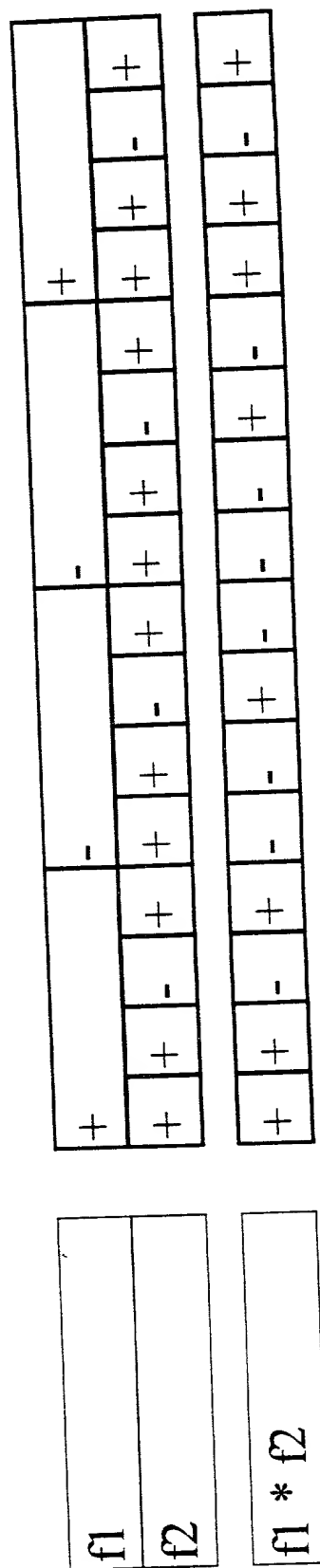
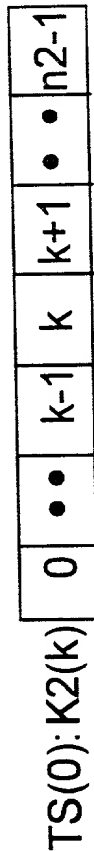
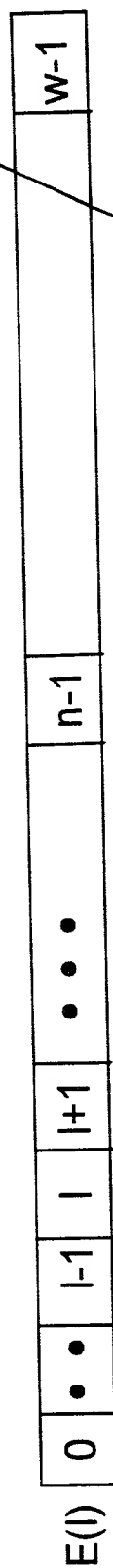


FIG 6

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$TS(n1*n2-1):$

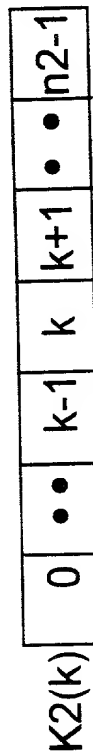


FIG 7

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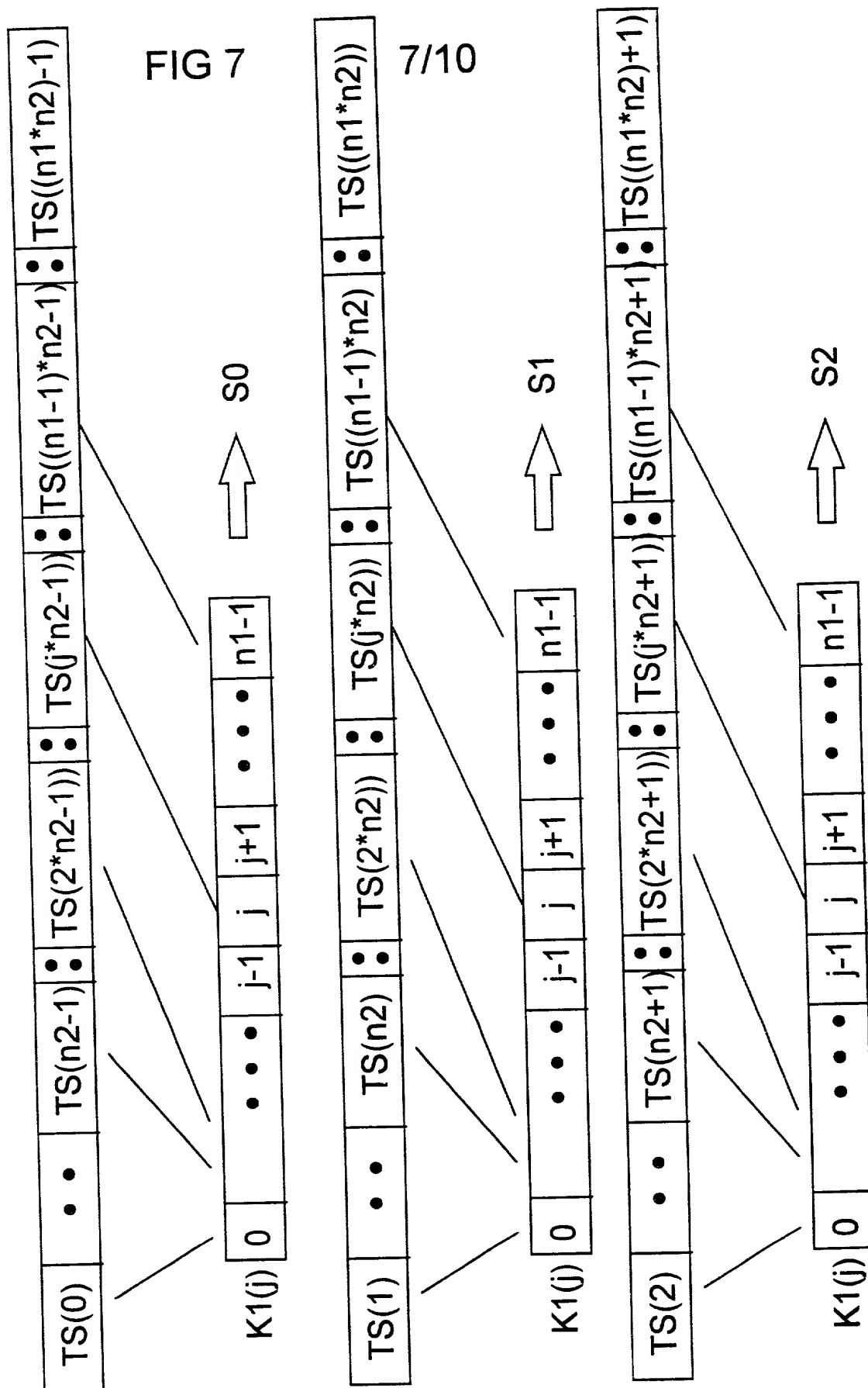


FIG 8

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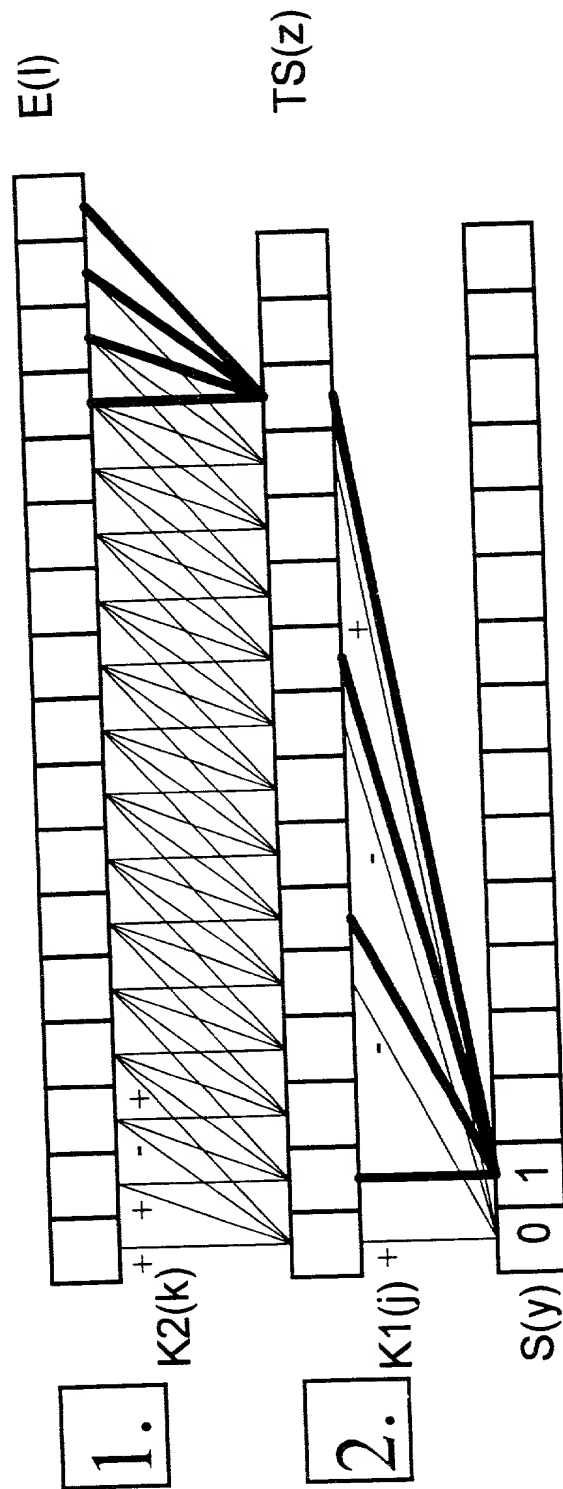


FIG 9

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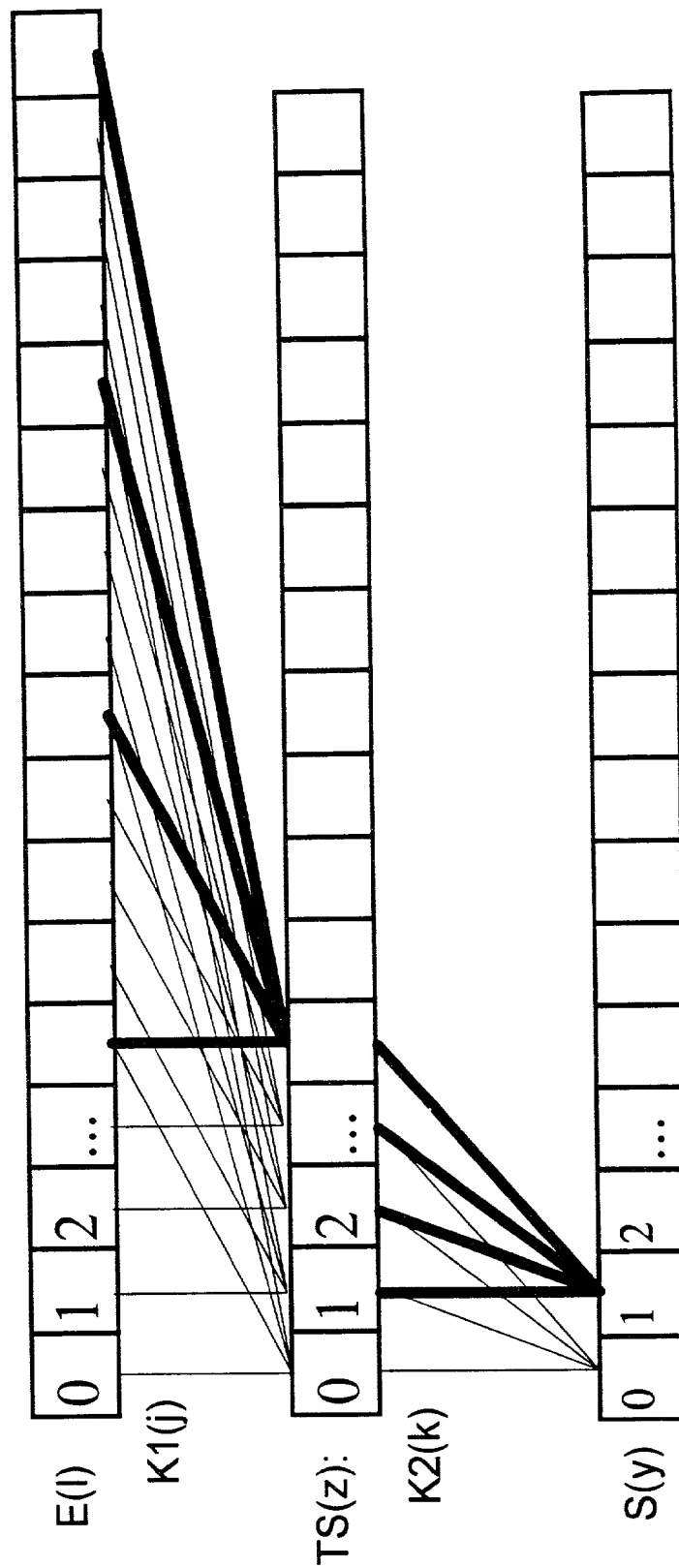
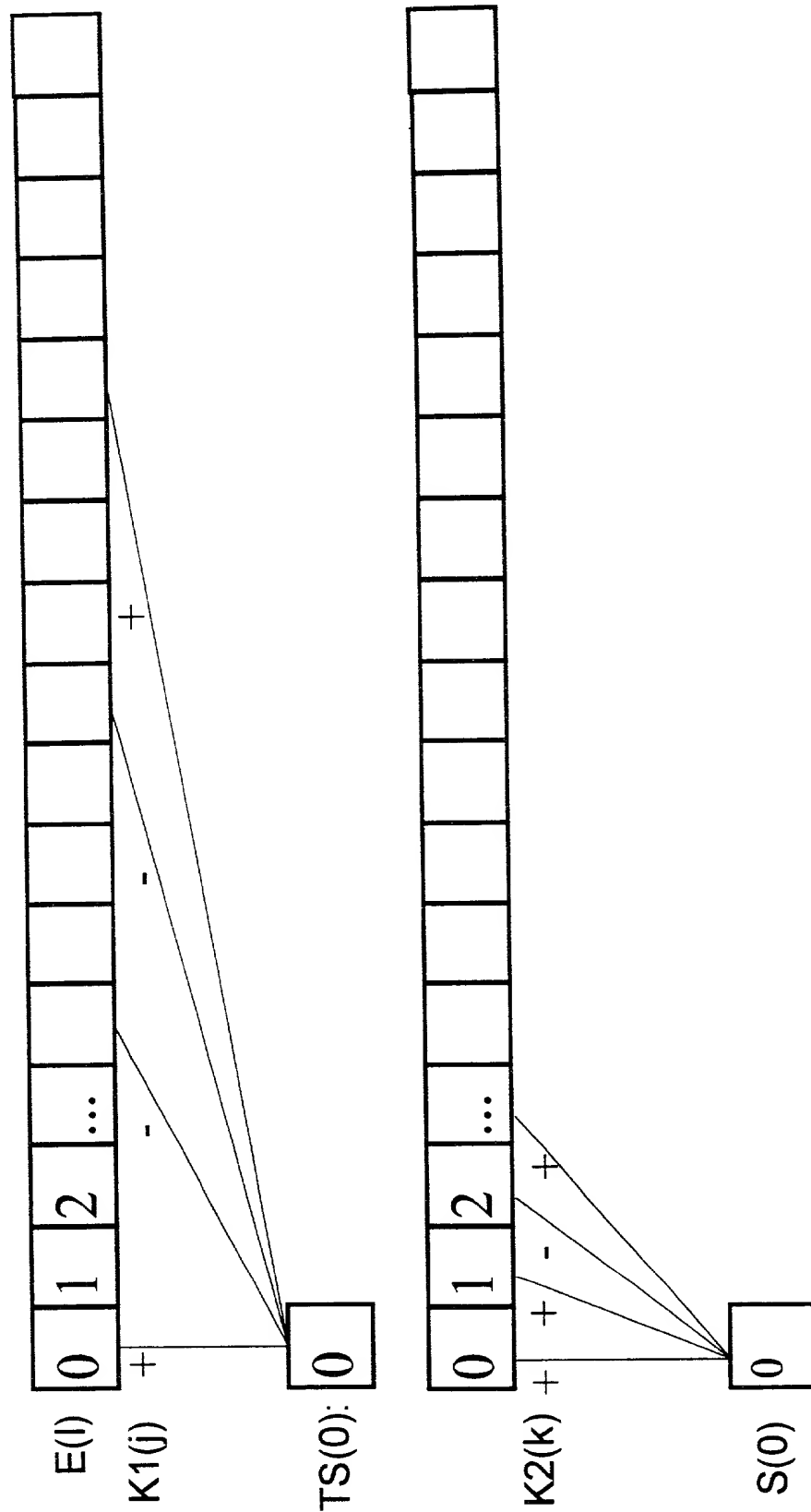


FIG 10

10/10



German Language Declaration

Prior foreign applications
Priorität beansprucht

Priority Claimed

198 40 974.5 Germany

(Number) (Country)
(Nummer) (Land)

08. September 1998

(Day Month Year Filed)
(Tag Monat Jahr eingereicht)



Yes
Ja



No
Nein

199 07 083.0 Germany

(Number) (Country)
(Nummer) (Land)

19. Februar 1999

(Day Month Year Filed)
(Tag Monat Jahr eingereicht)



Yes
Ja



No
Nein

(Number) (Country)
(Nummer) (Land)

(Day Month Year Filed)
(Tag Monat Jahr eingereicht)



Yes
Ja



No
Nein

Ich beanspruche hiermit gemäss Absatz 35 der Zivilprozessordnung der Vereinigten Staaten, Paragraph 120, den Vorzug aller unten aufgeführten Anmeldungen und falls der Gegenstand aus jedem Anspruch dieser Anmeldung nicht in einer früheren amerikanischen Patentanmeldung laut dem ersten Paragraphen des Absatzes 35 der Zivilprozessordnung der Vereinigten Staaten, Paragraph 122 offenbart ist, erkenne ich gemäss Absatz 37, Bundesgesetzbuch, Paragraph 1.56(a) meine Pflicht zur Offenbarung von Informationen an, die zwischen dem Anmeldedatum der früheren Anmeldung und dem nationalen oder PCT internationalen Anmeldedatum dieser Anmeldung bekannt geworden sind.

I hereby claim the benefit under Title 35, United States Code, §120 of any United States application(s) listed below and, insofar as the subject matter of each of the claims of this application is not disclosed in the prior United States application in the manner provided by the first paragraph of Title 35, United States Code, §122, I acknowledge the duty to disclose material information as defined in Title 37, Code of Federal Regulations, §1.56(a) which occurred between the filing date of the prior application and the national or PCT international filing date of this application

(Application Serial No.)
(Anmeldeseriennummer)

(Filing Date)
(Anmeldedatum)

(Status)
(patentiert, anhängig,
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(Status)
(patented, pending,
abandoned)

(Application Serial No.)
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(Filing Date)
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Declaration and Power of Attorney For Patent Application

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German Language Declaration

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I believe I am the original, first and sole inventor (if only one name is listed below) or an original, first and joint inventor (if plural names are listed below) of the subject matter which is claimed and for which a patent is sought on the invention entitled

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PCT Application No _____

and was amended on _____
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I acknowledge the duty to disclose information which is material to the examination of this application in accordance with Title 37, Code of Federal Regulations, §1 56(a)

I hereby claim foreign priority benefits under Title 35, United States Code, §119 of any foreign application(s) for patent or inventor's certificate listed below and have also identified below any foreign application for patent or inventor's certificate having a filing date before that of the application on which priority is claimed:

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POWER OF ATTORNEY: As a named inventor, I hereby appoint the following attorney(s) and/or agent(s) to prosecute this application and transact all business in the Patent and Trademark Office connected therewith (list name and registration number)

And I hereby appoint

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| Unterschrift des Erfinders | Datum | Inventor's signature | Date |
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(Supply similar information and signature for third and subsequent joint inventors).

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| Unterschrift des Erfinders | Datum | Inventor's signature | Date |
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| Staatsangehörigkeit | | Citizenship | |
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| | | | |
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